# **MIC2287C**



# 1.2MHz PWM White LED Driver with OVP in 2mm × 2mm MLF™ and Thin SOT-23

# **General Description**

The MIC2287C is a 1.2MHz pulse width modulated (PWM), boost-switching regulator that is optimized for constant-current, white LED driver applications. With a maximum output voltage of 37V and a switch current of over 500mA, the MIC2287C easily drives a string of up to 6 white LEDs in series, ensuring uniform brightness and eliminating several ballast resistors.

The MIC2287C implements a constant frequency, 1.2MHz PWM control scheme. The high frequency PWM operation saves board space by reducing external component sizes. The added benefit of the constant frequency PWM scheme as opposed to variable frequency topologies is much lower noise and input ripple injected back to the battery source.

To optimize efficiency, the feedback voltage is set to only 95mV. This reduces the power dissipation in the current set resistor and allows the lowest total output voltage, hence minimal current draw from the battery.

The MIC2287C is available with 3 levels of overvoltage protection, 15V, 24V, and 34V. This allows designers to choose the smallest possible external components with the appropriate voltage ratings for their applications.

The MIC2287C is available in low profile Thin SOT-23 5-lead and an 8-lead 2mm × 2mm MLF<sup>™</sup> package options. The MIC2287C has a junction temperature range of  $-40^{\circ}$ C to  $+105^{\circ}$ C.

All support documentation can be found on Micrel's web site at www.micrel.com.

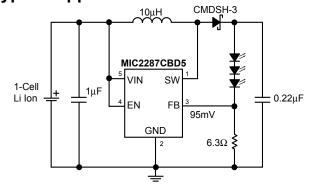
### **Features**

- 2.5V to 10V input voltage
- · Output voltage up to 34V
- Over 500mA switch current
- 1.2 MHz PWM operation
- · 95mV feedback voltage
- Output Overvoltage Protection (OVP)
- Options for 15V, 24V, and 34V OVP
- Over-temperature protection
- UVLO
- Low profile Thin SOT-23-5 package option
- 8-lead 2mm × 2mm MLF™ package option
- -40°C to +105°C junction temperature range
- For higher performance specifications see the MIC2287

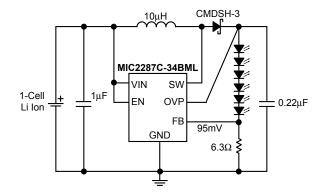
# **Applications**

- · White LED driver for backlighting:
  - Cell phones
  - PDAs
  - GPS systems
  - Digital cameras
  - MP3 players
  - IP phones
- LED flashlights

# **Typical Application**



3-Series White LED Driver in Thin SOT-23



6-Series White LED Driver with Output OVP in 2mm × 2mm MLF™-8

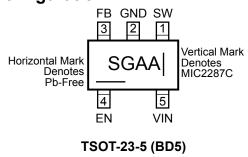
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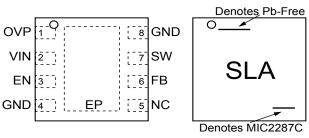
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**Ordering Information** 

Part Number	Marking Code	Overvoltage Protection	Junction Temp. Range	Package	Lead Finish	
MIC2287CBD5	SGAA	N/A	–40°C to 105°C	Thin SOT-23-5	Standard	
MIC2287CYD5	<u>S</u> GAA	N/A	–40°C to 105°C	Thin SOT-23-5	Lead Free	
MIC2287C-15BML	<u>SL</u> A	15V	–40°C to 105°C	2mm x 2mm MLF™	Standard	
MIC2287C-15YML	<u>SL</u> A	15V	–40°C to 105°C	2mm x 2mm MLF™	Lead Free	
MIC2287C-24BML	<u>SL</u> B	24V	–40°C to 105°C	2mm x 2mm MLF™	Standard	
MIC2287C-24YML	<u>SL</u> B	24V	–40°C to 105°C	2mm x 2mm MLF™	Lead Free	
MIC2287C-34BML	<u>SL</u> C	34V	-40°C to 105°C	2mm x 2mm MLF™	Standard	
MIC2287C-34YML	<u>SL</u> C	34V	–40°C to 105°C	2mm x 2mm MLF™	Lead Free	

# **Pin Configuration**





8-Pin MLF™ (BML) (Top View)

**Pin Description** 

Pin Number TSOT-23-5	Pin Number 2mm × 2mm MLF™	Pin Name	Pin Function	
1	7	SW	Switch node (Input): Internal power BIPOLAR collector.	
2		GND	Ground (Return): Ground.	
3	6	FB	Feedback (Input): Output voltage sense node. Connect the cathode of the LED to this pin. A resistor from this pin to ground sets the LED current.	
4	3	EN	Enable (Input): Logic high enables regulator. Logic low shuts down regulator.	
5	2	VIN	Supply (Input): 2.7V to 8V for internal circuitry.	
_	1	OVP	Overvoltage protection (Input): Connect to the output.	
_	4	AGND	Analog ground.	
_	8	PGND	Power ground.	
_	5	NC	No connect (no internal connection to die).	
_	EP	GND	Ground (Return): Exposed backside pad.	

# Absolute Maximum Ratings<sup>(1)</sup>

•	
Supply Voltage (V <sub>IN</sub> )	12V
Switch Voltage (V <sub>SW</sub> )	0.3V to 37V
Enable Pin Voltage (V <sub>EN</sub> )	–0.3 to V <sub>IN</sub>
FB Voltage (V <sub>FB</sub> )	6V
Switch Current (I <sub>SW</sub> )	2A
Ambient Storage Temperature (T <sub>S</sub> )	–65°C to +150°C
ESD Rating <sup>(3)</sup>	2kV

# Operating Ratings<sup>(2)</sup>

Supply Voltage (V <sub>IN</sub> )	2.5V to 10\
Junction Temperature Range (T <sub>J</sub> )	–40°C to +105°C
Package Thermal Impedance	
2mm × 2mmMLF™ (θ <sub>JA</sub> )	93°C/V
Thin SOT-23-5 (θ <sub>ΙΑ</sub> )	256°C/V

# Electrical Characteristics<sup>(4)</sup>

 $T_{A} = 25^{\circ}\text{C}, \ V_{IN} = V_{EN} = 3.6\text{V}, \ V_{OUT} = 10\text{V}, \ I_{OUT} = 10\text{mA}, \ \text{unless otherwise noted}. \ \textbf{Bold} \ \text{values indicate} \ -40^{\circ}\text{C} < T_{J} < 105^{\circ}\text{C}.$ 

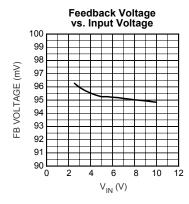
Symbol	Parameter	Condition	Min	Тур	Max	Units
$\overline{V_{IN}}$	Supply Voltage Range		2.5		10	V
V <sub>UVLO</sub>	Under Voltage Lockout		1.8	2.1	2.4	V
I <sub>VIN</sub>	Quiescent Current	V <sub>FB</sub> > 200mV, (not switching)		2.5	5	mA
I <sub>SD</sub>	Shutdown Current	V <sub>EN</sub> = 0V <sup>(5)</sup>		0.1	1	μΑ
$\overline{V_{FB}}$	Feedback Voltage	(±10%)	85	95	105	mV
I <sub>FB</sub>	Feedback Input Current	V <sub>FB</sub> = 95mV		-450		nA
	Line Regulation <sup>(6)</sup>	3V ≤ V <sub>IN</sub> ≤ 5V		0.5		%
	Load Regulation <sup>(6)</sup>	5mA ≤ I <sub>OUT</sub> ≤ 20mA		0.5		%
D <sub>MAX</sub>	Maximum Duty Cycle		85	90		%
I <sub>SW</sub>	Switch Current Limit			750		mA
$\overline{V_{SW}}$	Switch Saturation Voltage	I <sub>SW</sub> = 0.5A		450		mV
I <sub>SW</sub>	Switch Leakage Current	V <sub>EN</sub> = 0V, V <sub>SW</sub> = 10V		0.01	5	μΑ
$\overline{V_{EN}}$	Enable Threshold	TURN ON TURN OFF	1.5		0.4	V V
I <sub>EN</sub>	Enable Pin Current	V <sub>EN</sub> = 10V		20	40	μΑ
$f_{SW}$	Oscillator Frequency		1.05	1.2	1.35	MHz
V <sub>OVP</sub>	Overvoltage Protection	MIC2287CBML- 15 only MIC2287CBML- 24 only MIC2287CBML- 34 only	11.5 19 27	14 22.5 32	16.5 26 37	V V V
T <sub>J</sub>	Overtemperature Threshold Shutdown	Hysteresis		150 10		°C

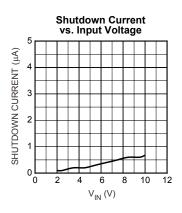
### Notes

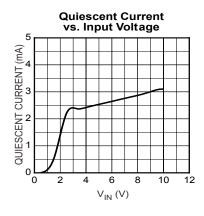
- 2. This device is not guaranteed to operate beyond its specified operating ratings.
- 3. Devices are inherently ESD sensitive. Handling precautions required. Human body model.
- 4. Specification for packaged product only.
- 5.  $I_{SD} = I_{VIN}$ .
- 6. Guaranteed by design

Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating
the device outside of its operating ratings. The maximum allowable power dissipation is a function of the maximum junction temperature, T<sub>J</sub>(max),
the junction-to-ambient thermal resistance, θ<sub>JA</sub>, and the ambient temperature, T<sub>A</sub>. The maximum allowable power dissipation will result in excessive
die temperature, and the regulator will go into thermal shutdown.

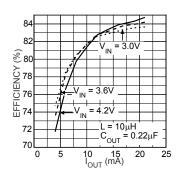
# **Typical Characteristics**



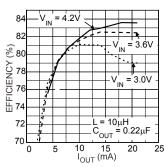


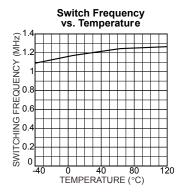




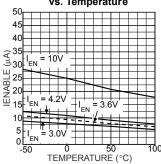


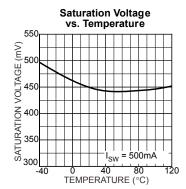


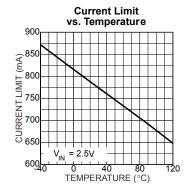


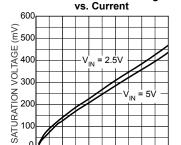








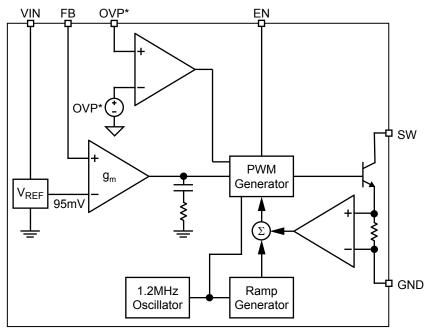




Switch Saturation Voltage

I<sub>SW</sub> (mA)

# **Functional Diagram**



\*OVP available on MLF™ package option only

MIC2287C Block Diagram

# **Functional Description**

The MIC2287C is a constant frequency, PWM current mode boost regulator. The block diagram is shown above. The MIC2287C is composed of an oscillator, slope compensation ramp generator, current amplifier,  $g_m$  error amplifier, PWM generator, and a 500mA bipolar output transistor. The oscillator generates a 1.2MHz clock. The clock's two functions are to trigger the PWM generator that turns on the output transistor and to reset the slope compensation ramp generator. The current amplifier is used to measure the switch current by amplifying the voltage signal from the internal sense resistor. The output of the current amplifier is summed with the output of the slope compensation ramp generator. This summed current-loop signal is fed to one of the inputs of the PWM generator.

The g<sub>m</sub> error amplifier measures the LED current through the

external sense resistor and amplifies the error between the detected signal and the 95mV reference voltage. The output of the  $g_m$  error amplifier provides the voltage-loop signal that is fed to the other input of the PWM generator. When the current-loop signal exceeds the voltage-loop signal, the PWM generator turns off the bipolar output transistor. The next clock period initiates the next switching cycle, maintaining the constant frequency current-mode PWM control. The LED is set by the feedback resistor:

$$I_{LED} = \frac{95mv}{R_{FB}}$$

The Enable pin shuts down the output switching and disables control circuitry to reduce input current-to-leakage levels. Enable pin input current is zero at zero volts.

# **External Component Selection**

capacitor values for various series-LED applications.

The MIC2287C can be used across a wide rage of applications. The table below shows recommended inductor and output

Series LEDs	L	Manufacturer	Min C <sub>OUT</sub>	Manufacturer
2	22µH	LQH32CN220K21 (Murata) NLC453232T-220K(TDK)	2.2µF	0805ZD225KAT(AVX) GRM40X5R225K10(Murata)
	15µH	LQH32CN150K21 (Murata) NLC453232T-150K(TDK)	1μF	0805ZD105KAT(AVX) GRM40X5R105K10(Murata)
	10µH	LQH32CN100K21 (Murata) NLC453232T-100K(TDK)	0.22µF	0805ZD224KAT(AVX) GRM40X5R224K10(Murata)
	6.8µH	LQH32CN6R8K21 (Murata) NLC453232T-6R8K(TDK)	0.22µF	0805ZD225KAT(AVX) GRM40X5R225K10(Murata)
	4.7µH	LQH32CN4R7K21 (Murata) NLC453232T-4R7K(TDK)	0.22µF	0805ZD224KAT(AVX) GRM40X5R224K10(Murata)
3	22µH	LQH43MN220K21 (Murata) NLC453232T-220K(TDK)	2.2µF	0805YD225MAT(AVX) GRM40X5R225K16(Murata)
	15µH	LQH43MN 150K21 (Murata) NLC453232T-150K(TDK)	1µF	0805YD105MAT(AVX) GRM40X5R105K16(Murata)
	10µH	LQH43MN 100K21 (Murata) NLC453232T-100K(TDK)	0.22µF	0805YD224MAT(AVX) GRM40X5R224K16(Murata)
	6.8µH	LQH43MN 6R8K21 (Murata) NLC453232T-6R8K(TDK)	0.22µF	0805YD224MAT(AVX) GRM40X5R224K16(Murata)
	4.7µH	LQH43MN 4R7K21 (Murata) NLC453232T-4R7K(TDK)	0.27µF	0805YD274MAT(AVX) GRM40X5R224K16(Murata)
4	22µH	LQH43MN220K21 (Murata) NLC453232T-220K(TDK)	1μF	0805YD105MAT(AVX) GRM40X5R105K25(Murata)
	15µH	LQH43MN 150K21 (Murata) NLC453232T-150K(TDK)	1μF	0805YD105MAT(AVX) GRM40X5R105K25(Murata)
	10µH	LQH43MN 100K21 (Murata) NLC453232T-100K(TDK)	0.27µF	0805YD274MAT(AVX) GRM40X5R274K25(Murata)
	6.8µH	LQH43MN 6R8K21 (Murata) NLC453232T-6R8K(TDK)	0.27µF	0805YD274MAT(AVX) GRM40X5R274K25(Murata)
	4.7µH	LQH43MN 4R7K21 (Murata) NLC453232T-4R7K(TDK)	0.27µF	0805YD274MAT(AVX) GRM40X5R274K25(Murata)
5, 6	22µH	LQH43MN220K21 (Murata) NLC453232T-220K(TDK)	0.22µF	08053D224MAT(AVX) GRM40X5R224K25(Murata)
	15µH	LQH43MN 150K21 (Murata) NLC453232T-150K(TDK)	0.22µF	08053D224MAT(AVX) GRM40X5R224K25(Murata)
	10µH	LQH43MN 100K21 (Murata) NLC453232T-100K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)
	6.8µH	LQH43MN 6R8K21 (Murata) NLC453232T-6R8K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)
	4.7µH	LQH43MN 4R7K21 (Murata) NLC453232T-4R7K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)
7, 8	22µH	LQH43MN220K21 (Murata) NLC453232T-220K(TDK)	0.22µF	08053D224MAT(AVX) GRM40X5R224K25(Murata)
	15µH	LQH43MN 150K21 (Murata) NLC453232T-150K(TDK)	0.22µF	08053D224MAT(AVX) GRM40X5R224K25(Murata)
	10µH	LQH43MN 100K21 (Murata) NLC453232T-100K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)
	6.8µH	LQH43MN 6R8K21 (Murata) NLC453232T-6R8K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)
	4.7µH	LQH43MN 4R7K21 (Murata) NLC453232T-4R7K(TDK)	0.27µF	08053D274MAT(AVX) GRM40X5R274K25(Murata)

### **Dimming Control**

There are two techniques for dimming control. One is PWM dimming, and the other is continuous dimming.

- 1. PWM dimming control is implemented by applying a PWM signal on EN pin as shown in Figure 1. The MIC2287C is turned on and off by the PWM signal. With this method, the LEDs operate with either zero or full current. The average LED current is increased proportionally to the duty-cycle of the PWM signal. This technique has high-efficiency because the IC and the LEDs consume no current during the off cycle of the PWM signal. Typical PWM frequency should be between 100Hz and 10kHz.
- 2. Continuous dimming control is implemented by applying a DC control voltage to the FB pin of the MIC2287C through a series resistor as shown in Figure 2. The LED intensity (current) can be dynamically varied applying a DC voltage to the FB pin. The DC voltage can come from a DAC signal, or a filtered PWM signal. The advantage of this approach is a high frequency PWM signal (>10kHz) that can be used to control LED intensity.

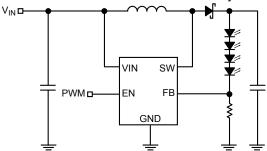


Figure 1. PWM Dimming Method

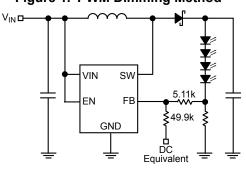


Figure 2. Continuous Dimming

### **Open-Circuit Protection**

If the LEDs are disconnected from the circuit, or in case an LED fails open, the sense resistor will pull the FB pin to ground. This will cause the MIC2287C to switch with a high duty-cycle, resulting in output overvoltage. This may cause the SW pin voltage to exceed its maximum voltage rating, possibly damaging the IC and the external components. To ensure the highest level of protection, the MIC2287C has 3 product options in the 8-lead 2mm × 2mm MLF™ with overvoltage protection (OVP). The extra pins of the 8-lead 2mm × 2mm MLF™ package allow a dedicated OVP monitor with options for 15V, 24V, or 34V (see Figure 3). The reason for

the three OVP levels is to let users choose the suitable level of OVP for their application. For example, a 3-LED application would typically see an output voltage of no more than 12V, so a 15V OVP option would offer a suitable level of protection. This allows the user to select the output diode and capacitor with the lowest voltage ratings, as well as smallest size and lowest cost. The OVP will clamp the output voltage to within the specified limits. For the Thin SOT-23-5 package, an OVP pin is not available. An external zener diode can be connected from the output of the converter to FB pin as shown in Figure 4. to implement similar protection.

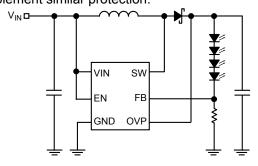


Figure 3. MLF™ Package OVP Circuit

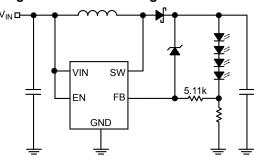


Figure 4. Thin SOT-23 Package OVP Circuit

### Start-Up and Inrush Current

During start-up, inrush current of approximately double the nominal current flows to set up the inductor current and the voltage on the output capacitor. If the inrush current needs to be limited, a soft-start circuit similar to Figure 5 could be implemented. The soft-start capacitor,  $C_{\rm SS}$ , provides overdrive to the FB pin at start-up, resulting in gradual increase of switch duty cycle and limited inrush current.

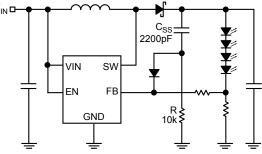
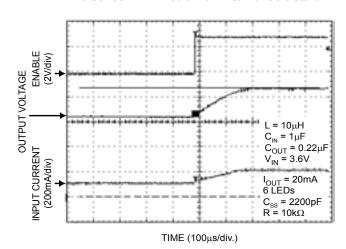


Figure 5. Soft-Start Circuit

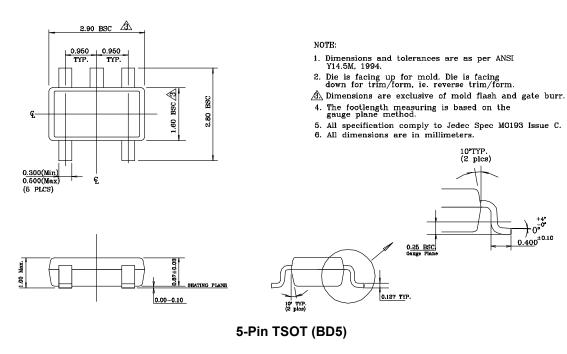
### 6-Series LED Circuit without External Soft-Start

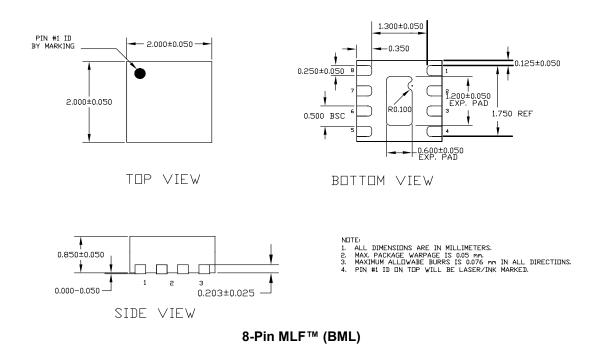
# CONTPUT VOLTAGE INPUT CURRENT COMP = 1μF COUT = 0.22μF V<sub>IN</sub> = 3.6V IOUT = 20mA 6 LEDs TIME (100μs/div.)

### 6-Series LED Circuit with External Soft-Start



# Package Information





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