# LVDS CLOCK MULTIPLIER FOR VIDEO APPLICATIONS

# ICS874S336

## General Description

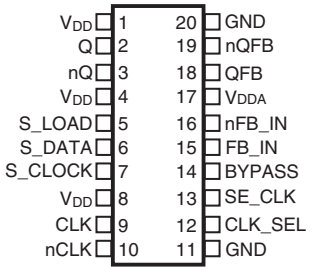
The ICS874S336 is a high performance, 1-to-1, Differential-to-LVDS Clock Multiplier and is a member of the HiPerClocksS™family of High Performance Clock Solutions from IDT. The CLK/ nCLK input pair can accept most standard differential input levels. The ICS874S336 has a fully integrated PLL along with frequency configurable outputs. An external feedback output regenerates clocks with “zero delay”.

The ICS874S336 has multiple divide combinations designed to work with the most common video rates used in professional video systems.

## Features

- One LVDS differential output pair, plus one LVDS feedback output pair
- One differential clock input pair CLK/nCLK can accept the following differential input levels: LVPECL, LVDS, LVHSTL, HCSL, SSTL
- Input Frequency Range: 14MHz to 17MHz
- Maximum Output Frequency: 204MHz
- VCO range: 1.2GHz – 2GHz
- Cycle-to-cycle jitter: TBD
- 3.3V operating supply voltage
- Low PLL bandwidth allows for better jitter attenuation
- 0°C to 70°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages

## Pin Assignment

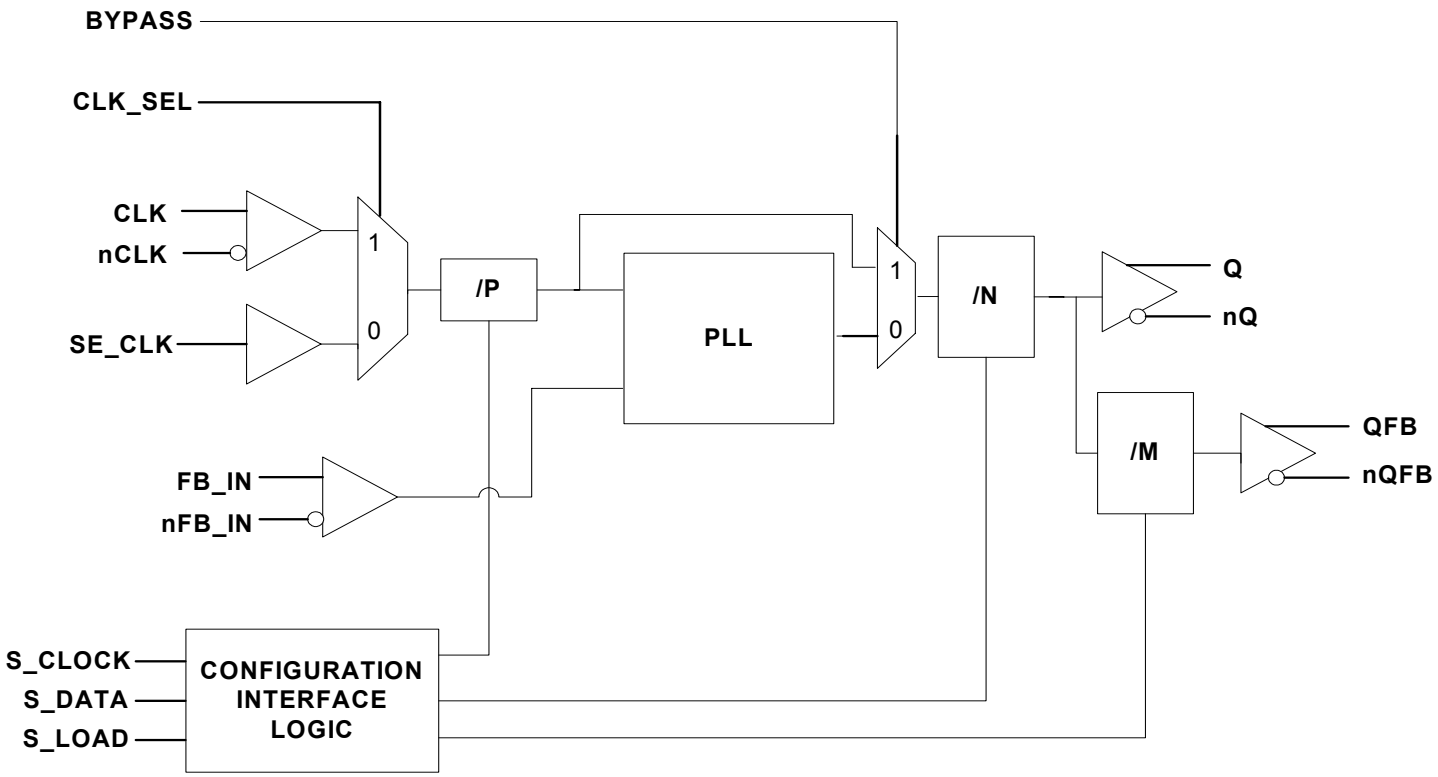



**ICS874S336I**
**20-Lead TSSOP**
**6.5mm x 4.4mm x 0.925mm**
**package body**

**G Package**
**Top View**

The Preliminary Information presented herein represents a product in pre-production. The noted characteristics are based on initial product characterization and/or qualification. Integrated Device Technology, Incorporated (IDT) reserves the right to change any circuitry or specifications without notice.

## Block Diagram



## Functional Description

The ICS874S336 features a fully integrated PLL and therefore requires no external components for setting the loop bandwidth. The VCO of the PLL operates over a range of 1.2GHz to 2GHz. The output of the M divider is also applied to the phase detector.

The phase detector and the M divider force the VCO output frequency to be M times the reference frequency by adjusting the VCO control voltage. Note that for some values of M (either too high or too low), the PLL will not achieve lock. The output of the VCO is scaled by a divider prior to being sent to each of the LVPECL output buffers. The divider provides a 50% output duty cycle.

The relationship between the VCO frequency, the input frequency and the M divider is defined as follows:

$$fVCO = \frac{fIN \times M \times N}{P} \times 2$$

The M, N, and P values used to obtain the proper video frequencies are found in Table 3B, *Programmable VCO Frequency Function Table.* The actual data bits can be found in Tables 3C, 3D and 3E.

Serial operation occurs when S_LOAD is LOW. The shift register is loaded by sampling the S_DATA bits with the rising edge of S_CLOCK. The contents of the shift register are loaded into the M, N and P dividers when S_LOAD transitions from LOW-to-HIGH. The divide values are latched on the HIGH-to-LOW transition of S_LOAD. If S_LOAD is held HIGH, data at the S_DATA input is passed directly to the dividers on each rising edge of S_CLOCK. The serial mode can be used to program the M, N and P bits.

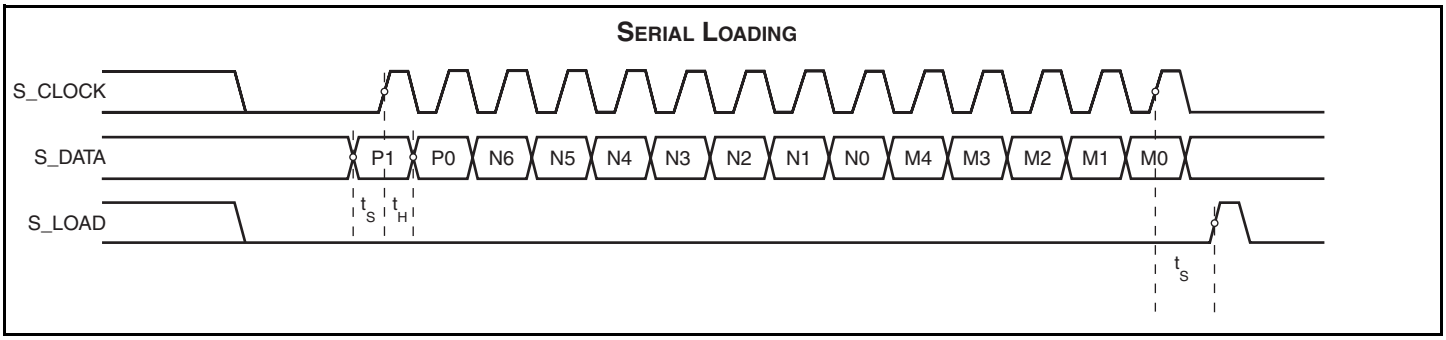



**Figure 1. Serial Load Operation**

## Table 1. Pin Descriptions

| Number | Name | Type | | Description |
|---|---|---|---|---|
| 1, 4, 8 | $V_{DD}$ | Power | | Core supply pins. |
| 2, 3 | Q, nQ | Output | | Differential output pair. LVDS interface levels. |
| 5 | S_LOAD | Input | Pulldown | Controls transition of data from shift register into the dividers. LVCMOS/LVTTL interface levels. |
| 6 | S_DATA | Input | Pulldown | Shift register serial input. Data sampled on the rising edge of S_CLOCK. LVCMOS/LVTTL interface levels. |
| 7 | S_CLOCK | Input | Pulldown | Clocks in serial data present at S_DATA input into the shift register on the rising edge of S_CLOCK. LVCMOS/LVTTL interface levels. |
| 9 | CLK | Input | Pulldown | Non-inverting differential clock input. |
| 10 | nCLK | Input | Pullup/Pulldown | Inverting differential clock input. $V_{DD}/2$ default when left floating. |
| 11, 20 | GND | Power | | Negative supply pin. |
| 12 | CLK_SEL | Input | Pullup | Selects the reference clock. When LOW selects SE_CLK as the clock source. When HIGH selects CLK, nCLK as the clock source. LVCMOS/LVTTL interface levels. |
| 13 | SE_CLK | Input | Pulldown | Single-ended clock input. LVCMOS/LVTTL interface levels. |
| 14 | BYPASS | Input | Pulldown | Selects between the PLL and reference clock as the input to the dividers.When LOW, selects PLL. When HIGH, selects reference clock. LVCMOS/LVTTL interface levels. |
| 15 | FB_IN | Input | Pulldown | Non-inverting differential clock input. |
| 16 | nFB_IN | Input | Pullup/Pulldown | Inverting differential clock input. $V_{DD}/2$ default when left floating. |
| 17 | $V_{DDA}$ | Power | | Analog supply pin. |
| 18, 19 | QFB, nQFB | Output | | Differential output pair. LVDS interface levels. |

NOTE: *Pullup and Pulldown* refer to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

## Table 2. Pin Characteristics

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|
| $C_{IN}$ | Input Capacitance | | | 4 | | pF |
| $R_{PULLUP}$ | Input Pullup Resistor | | | 51 | | kΩ |
| $R_{PULLDOWN}$ | Input Pulldown Resistor | | | 51 | | kΩ |

# Function Tables

**Table 3A. Parallel and Serial Mode Function Table**

| Inputs | | | |
|---|---|---|---|
| **S_LOAD** | **S_CLOCK** | **S_DATA** | **Conditions** |
| L | X | X | Data is latched into input registers and remains loaded until next LOW transition or until a serial event occurs. |
| L | ↑ | Data | Shift register is loaded with data on S_DATA on each rising edge of S_CLOCK. |
| ↑ | L | Data | Contents of the shift register are passed to the M, N and P dividers. |
| ↓ | L | Data | M, N and P divider values are latched. |
| L | X | X | Serial input do not affect shift registers. |
| H | ↑ | Data | S_DATA passed directly to M, N and P dividers as it is clocked. |

NOTE: L = LOW
H = HIGH
X = Don't care
↑ = Rising edge transition
↓ = Falling edge transition

**Table 3B. Device Configuration Table**

| Input Frequency (MHz) | | P Divide Value | N Divide Value | M Divide Value | Output Frequency (MHz) | |
|---|---|---|---|---|---|---|
| **Min** | **Max** | | | | **Min** | **Max** |
| 14 | 17 | 1 | 10 | 12 | 168 | 204 |
| 14 | 17 | 1 | 10 | 10 | 140 | 170 |
| 14 | 17 | 2 | 12 | 17 | 119 | 144.5 |
| 14 | 17 | 4 | 14 | 28 | 98 | 119 |
| 14 | 17 | 4 | 16 | 24 | 84 | 102 |
| 14 | 17 | 4 | 20 | 20 | 70 | 85 |
| 14 | 17 | 4 | 22 | 17 | 59.5 | 72.25 |
| 14 | 17 | 4 | 28 | 14 | 49 | 59.5 |
| 14 | 17 | 4 | 32 | 12 | 42 | 51 |
| 14 | 17 | 8 | 38 | 20 | 35 | 42.5 |
| 14 | 17 | 8 | 46 | 17 | 29.75 | 36.125 |
| 14 | 17 | 8 | 56 | 14 | 24.5 | 29.75 |
| 14 | 17 | 8 | 64 | 12 | 21 | 25.5 |
| 14 | 17 | 8 | 80 | 10 | 17.5 | 21.25 |
| 14 | 17 | 8 | 100 | 8 | 14 | 17 |
| 14 | 17 | 8 | 110 | 7 | 12.25 | 14.875 |
| 14 | 17 | 8 | 130 | 6 | 10.5 | 12.75 |
| 14 | 17 | 8 | 160 | 5 | 8.75 | 10.6 |

**Table 3C. Pre-Divider (P) Configuration Table**

| P Divide | P1 | P0 |
|---|---|---|
| 1 | 0 | 0 |
| 2 | 0 | 1 |
| 4 | 1 | 0 |
| 8 | 1 | 1 |

**Table 3D. Output Divider (N) Configuration Table**

| N Divide | N6 | N5 | N4 | N3 | N2 | N1 | N0 |
|---|---|---|---|---|---|---|---|
| 10 | 0 | 0 | 0 | 0 | 1 | 0 | 1 |
| 12 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |
| 14 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |
| 16 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 20 | 0 | 0 | 0 | 1 | 0 | 1 | 0 |
| 22 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |
| 28 | 0 | 0 | 0 | 1 | 1 | 1 | 0 |
| 32 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| 38 | 0 | 0 | 1 | 0 | 0 | 1 | 1 |
| 46 | 0 | 0 | 1 | 0 | 1 | 1 | 1 |
| 56 | 0 | 0 | 1 | 1 | 1 | 0 | 0 |
| 64 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| 80 | 0 | 1 | 0 | 1 | 0 | 0 | 0 |
| 100 | 0 | 1 | 1 | 0 | 0 | 1 | 0 |
| 110 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |
| 130 | 1 | 0 | 0 | 0 | 0 | 0 | 1 |
| 160 | 1 | 0 | 1 | 0 | 0 | 0 | 0 |

**Table 3E. Feedback Divider (M) Configuration Table**

| M Divide | M4 | M3 | M2 | M1 | M0 |
|---|---|---|---|---|---|
| 5 | 0 | 0 | 1 | 0 | 1 |
| 6 | 0 | 0 | 1 | 1 | 0 |
| 7 | 0 | 0 | 1 | 1 | 1 |
| 8 | 0 | 1 | 0 | 0 | 0 |
| 10 | 0 | 1 | 0 | 1 | 0 |
| 12 | 0 | 1 | 1 | 0 | 0 |
| 14 | 0 | 1 | 1 | 1 | 0 |
| 17 | 1 | 0 | 0 | 0 | 1 |
| 20 | 1 | 0 | 1 | 0 | 0 |
| 24 | 1 | 1 | 0 | 0 | 0 |
| 28 | 1 | 1 | 1 | 0 | 0 |

## Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

| Item | Rating |
|---|---|
| Supply Voltage, $V_{DD}$ | 4.6V |
| Inputs, $V_I$ | -0.5V to $V_{DD}$ + 0.5V |
| Outputs, $I_O$ (LVDS)<br>Continuos Current<br>Surge Current<br><br>Outputs, $I_O$ (LVDS)<br>Continuos Current<br>Surge Current | <br>10mA<br>15mA<br><br><br>50mA<br>100mA |
| Package Thermal Impedance, $\theta_{JA}$ | 87.2°C/W (0 mps) |
| Storage Temperature, $T_{STG}$ | -65°C to 150°C |

## DC Electrical Characteristics

**Table 4A. LVDS Power Supply DC Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|
| $V_{DD}$ | Positive Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| $V_{DDA}$ | Analog Supply Voltage | | $V_{DD}$ – 0.15 | 3.3 | $V_{DD}$ | V |
| $I_{DD}$ | Power Supply Current | | | 115 | | mA |
| $I_{DDA}$ | Analog Supply Current | | | 15 | | mA |

**Table 4B. LVCMOS/LVTTL DC Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

| Symbol | Parameter | | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|---|
| $V_{IH}$ | Input High Voltage | | | 2 | | $V_{DD}$ + 0.3 | V |
| $V_{IL}$ | Input Low Voltage | | | -0.3 | | 0.8 | V |
| $I_{IH}$ | Input High Current | SE_CLK, BYPASS, S_CLOCK, S_DATA, S_LOAD | $V_{DD}$ = $V_{IN}$ = 3.465V | | | 150 | µA |
| | | CLK_SEL | $V_{DD}$ = $V_{IN}$ = 3.465V | | | 5 | µA |
| $I_{IL}$ | Input Low Current | SE_CLK, BYPASS, S_CLOCK, S_DATA, S_LOAD | $V_{DD}$ = 3.465V, $V_{IN}$ = 0V | -5 | | | µA |
| | | CLK_SEL | $V_{DD}$ = 3.465V, $V_{IN}$ = 0V | -150 | | | µA |

**Table 4C. Differential DC Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

| Symbol | Parameter | | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|---|
| $I_{IH}$ | Input High Current | CLK/nCLK, FB_IN/nFB_IN | $V_{DD} = V_{IN}$ = 3.465V | | | 150 | µA |
| $I_{IL}$ | Input Low Current | CLK, FB_IN | $V_{DD}$ = 3.465V, $V_{IN}$ = 0V | -5 | | | µA |
| | | nCLK, nFB_IN | $V_{DD}$ = 3.465V, $V_{IN}$ = 0V | -150 | | | µA |
| $V_{PP}$ | Peak-to-Peak Voltage; NOTE 1 | | | 0.15 | | 1.3 | V |
| $V_{CMR}$ | Common Mode Input Voltage; NOTE 1, 2 | | | GND + 0.5 | | $V_{DD}$ – 0.85 | V |

NOTE 1: $V_{IL}$ should not be less than -0.3V.
NOTE 2: Common mode input voltage is defined as $V_{IH}$.

**Table 4D. LVDS DC Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|
| $V_{OD}$ | Differential Output Voltage | | | 370 | | mV |
| $\Delta V_{OD}$ | $V_{OD}$ Magnitude Change | | | 50 | | mV |
| $V_{OS}$ | Offset Voltage | | | 1.22 | | V |
| $\Delta V_{OS}$ | $V_{OS}$ Magnitude Change | | | 50 | | mV |

**Table 5. Input Frequency Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

| Symbol | Parameter | | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|---|
| $f_{IN}$ | Input Frequency | CLK/nCLK, SE_CLK; NOTE 1 | | 14 | | 17 | MHz |
| | | S_CLOCK | | | | 10 | MHz |

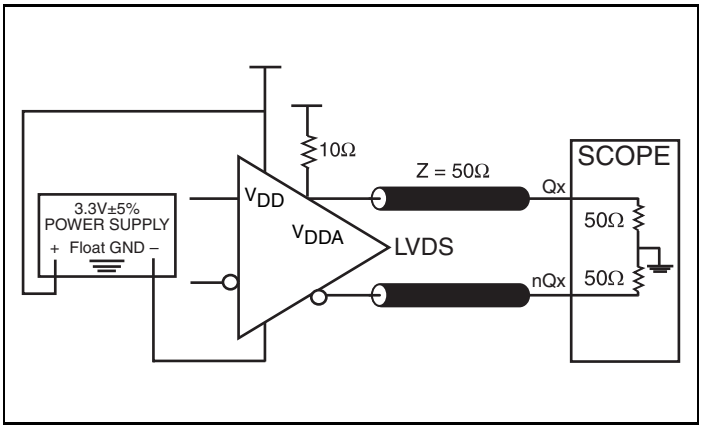
NOTE 1: For the CLK/nCLK and SE_CLK frequency range, the M value must be set for the VCO to operate within the TBD MHz to TBD MHz range.

**Table 6. AC Characteristics,** $V_{DD}$ = 3.3V ± 5%, $T_A$ = 0°C to 70°C

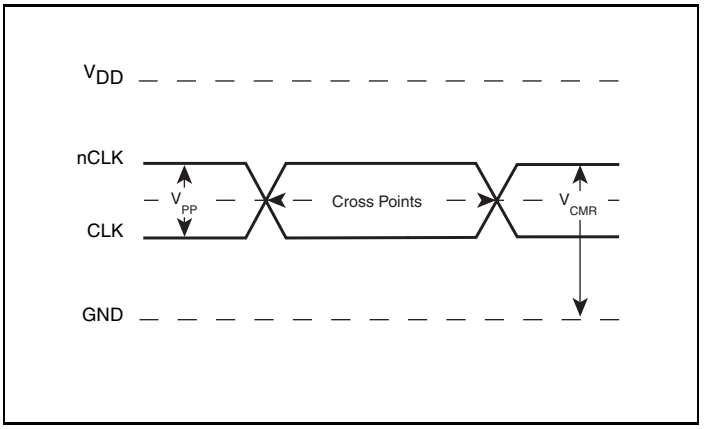
| Parameter | Symbol | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|---|---|---|---|---|
| $f_{MAX}$ | Output Frequency | | 8.75 | | 204 | MHz |
| *t*jit(cc) | Cycle-to-Cycle Jitter; NOTE 1 | | | TBD | | ps |
| *t*(Ø) | Static Phase Offset; NOTE 1 | | | TBD | | ps |
| *t*jit(per) | Period Jitter, RMS; NOTE 1 | | | TBD | | ps |
| $t_R$ / $t_F$ | Output Rise/Fall Time | 20% to 80% | | 270 | | ps |
| odc | Output Duty Cycle | | | 50 | | % |

NOTE 1: This parameter is defined in accordance with JEDEC Standard 65.

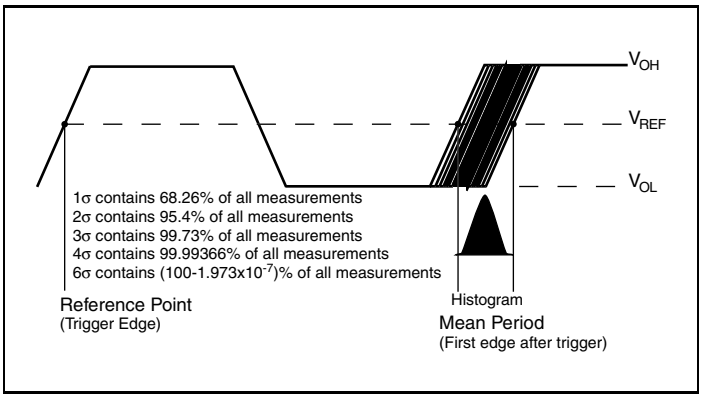
# Parameter Measurement Information



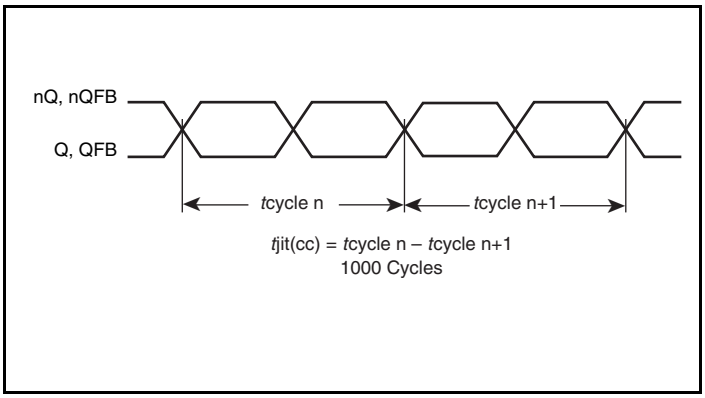

**3.3V LVDS Output Load AC Test Circuit**



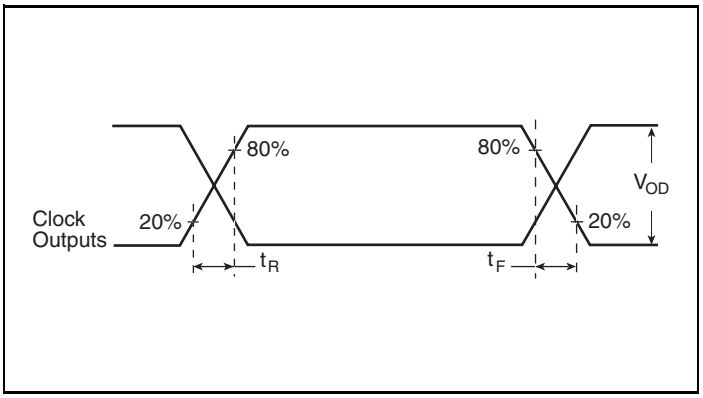

**Differential Input Level**



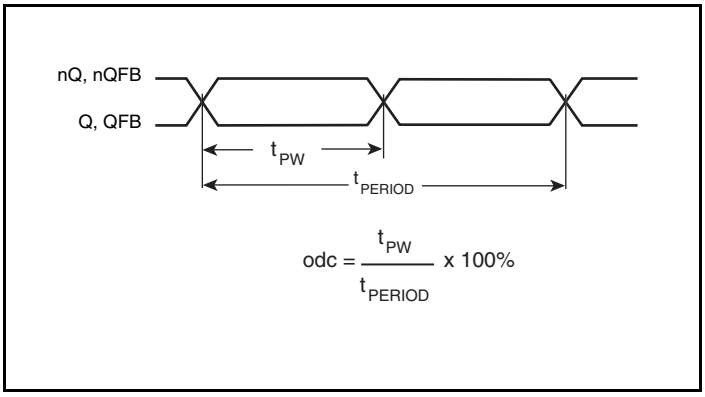

**Period Jitter, RMS**




**Cycle-to-Cycle Jitter**

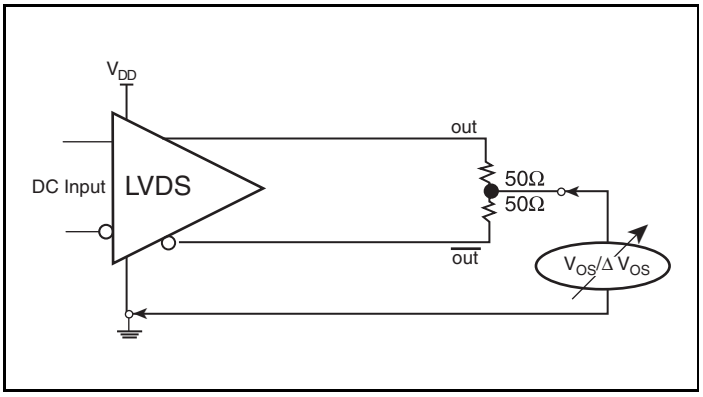



**Output Rise/Fall Time**

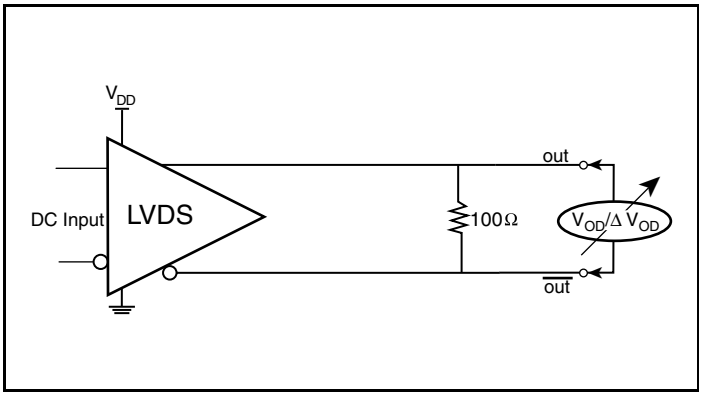



**Output Duty Cycle/Pulse Width/Period**

## Parameter Measurement Information, continued




**Offset Voltage Setup**




**Differential Output Voltage Setup**

## Application Information

### Wiring the Differential Input to Accept Single-Ended Levels

*Figure 2* shows how the differential input can be wired to accept single-ended levels. The reference voltage V_REF = $V_{DD}$/2 is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio of R1 and R2 might need to be adjusted to position the V_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and $V_{DD}$ = 3.3V, V_REF should be 1.25V and R2/R1 = 0.609.

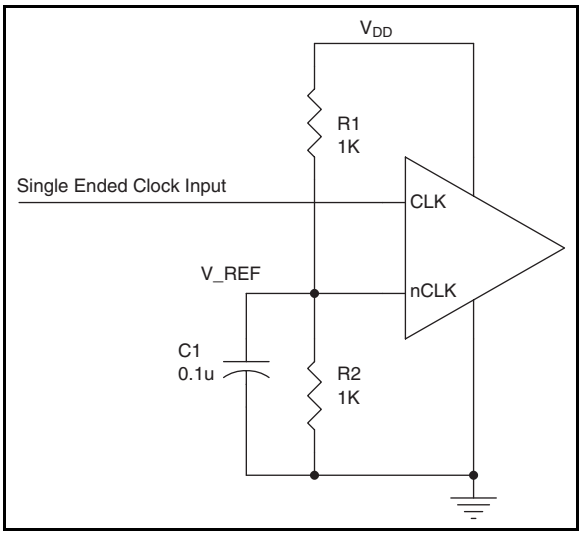



**Figure 2. Single-Ended Signal Driving Differential Input**

## Differential Clock Input Interface

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both $V_{SWING}$ and $V_{OH}$ must meet the $V_{PP}$ and $V_{CMR}$ input requirements. *Figures 3A to 3F* show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in Figure 3A, the input termination applies for IDT HiPerClockS open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

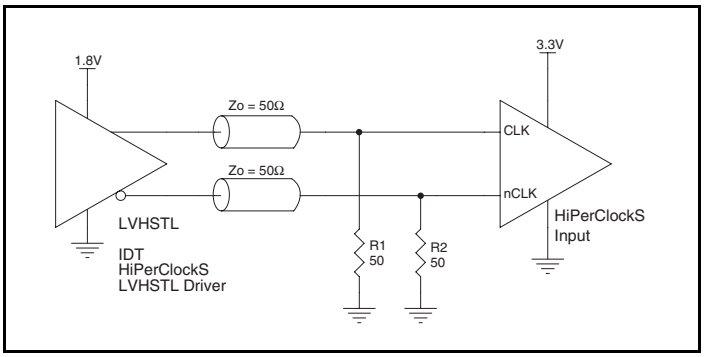



**Figure 3A. HiPerClockS CLK/nCLK Input Driven by an IDT Open Emitter HiPerClockS LVHSTL Driver**

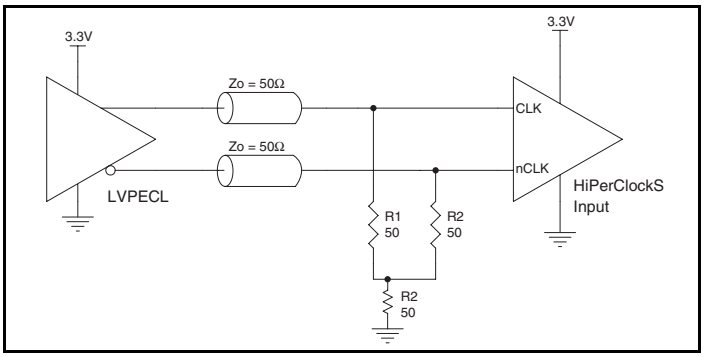



**Figure 3B. HiPerClockS CLK/nCLK Input Driven by a 3.3V LVPECL Driver**

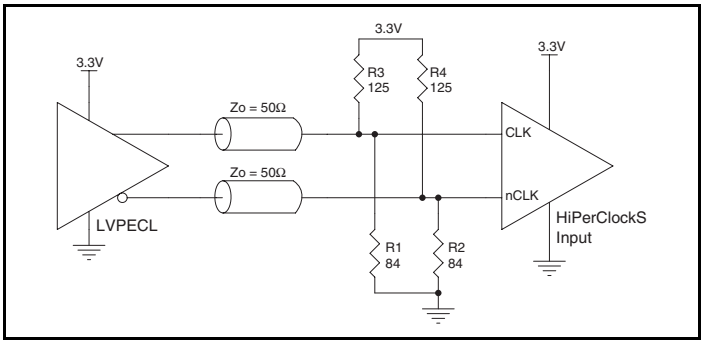



**Figure 3C. HiPerClockS CLK/nCLK Input Driven by a 3.3V LVPECL Driver**

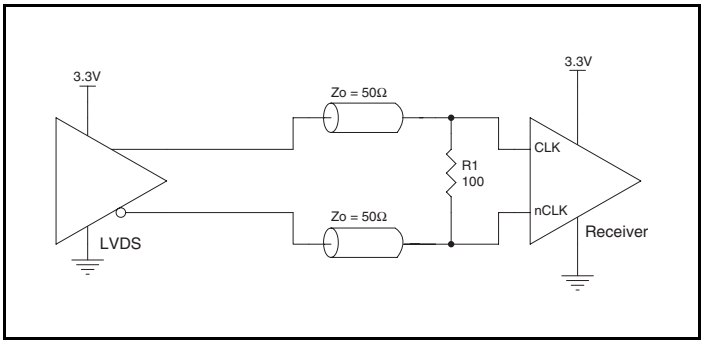



**Figure 3D. HiPerClockS CLK/nCLK Input Driven by a 3.3V LVDS Driver**

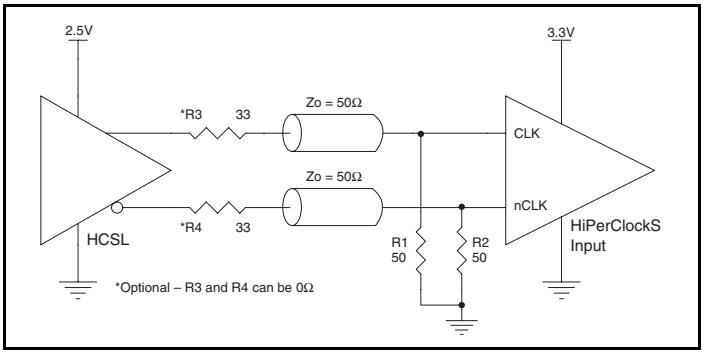



**Figure 3E. HiPerClockS CLK/nCLK Input Driven by a 3.3V HCSL Driver**

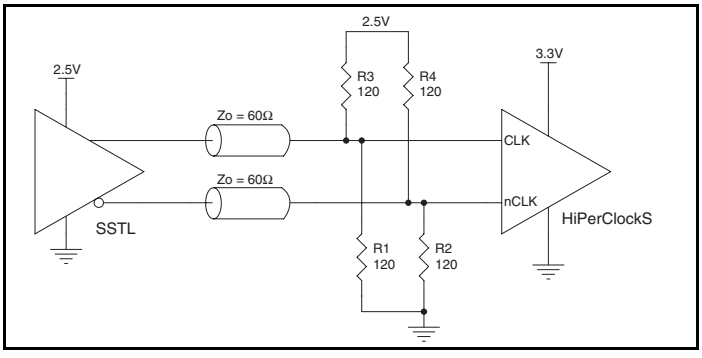



**Figure 3F. HiPerClockS CLK/nCLK Input Driven by a 2.5V SSTL Driver**

## Power Supply Filtering Technique

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS874S336 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $V_{DD}$ and $V_{DDA}$ should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. *Figure 4* illustrates how a 10Ω resistor along with a 10μF and a 0.01μF bypass capacitor should be connected to each $V_{DDA}$ pin.

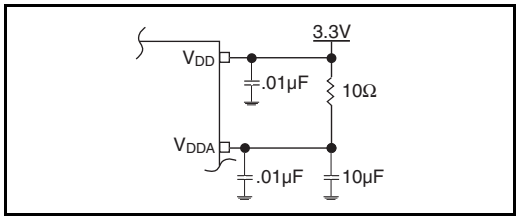



**Figure 4. Power Supply Filtering**

## Recommendations for Unused Input and Output Pins

### Inputs:

#### CLK/nCLK Inputs

For applications not requiring the use of the differential input, both CLK and nCLK can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from CLK to ground.

#### SE_CLK Input

For applications not requiring the use of a clock input, it can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from the SE_CLK input to ground.

#### LVCMOS Control Pins

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

### Outputs:

#### LVDS Outputs

All unused LVDS output pairs can be either left floating or terminated with 100Ω across. If they are left floating, there should be no trace attached.

## 3.3V LVDS Driver Termination

A general LVDS interface is shown in *Figure 5.* In a 100Ω differential transmission line environment, LVDS drivers require a matched load termination of 100Ω across near the receiver input.

For a multiple LVDS outputs buffer, if only partial outputs are used, it is recommended to terminate the unused outputs.

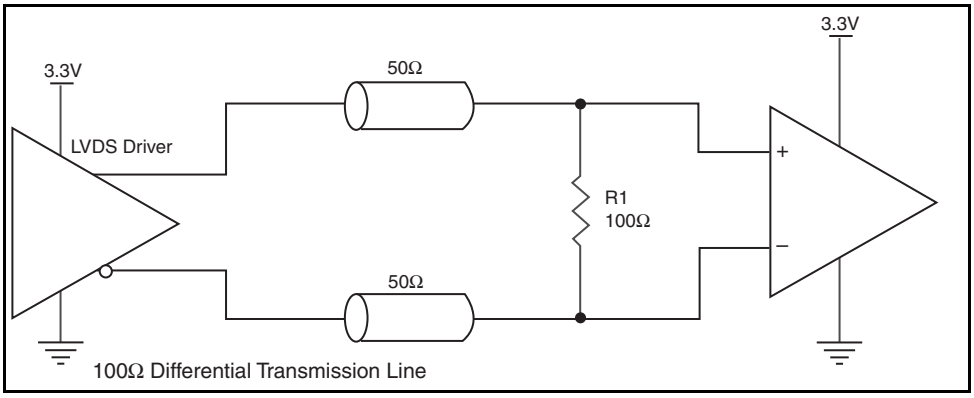



**Figure 5. Typical LVDS Driver Termination**

## Schematic Example

*Figure 6* shows an example of ICS874S336 application schematic. In this example, the device is operated at $V_{DD}$ = 3.3V. The decoupling capacitors should be located as close as possible to the power pin. Two examples of LVDS terminations are shown in this schematic. The input is driven either by a 3.3V LVPECL driver or a 3.3V LVCMOS. .

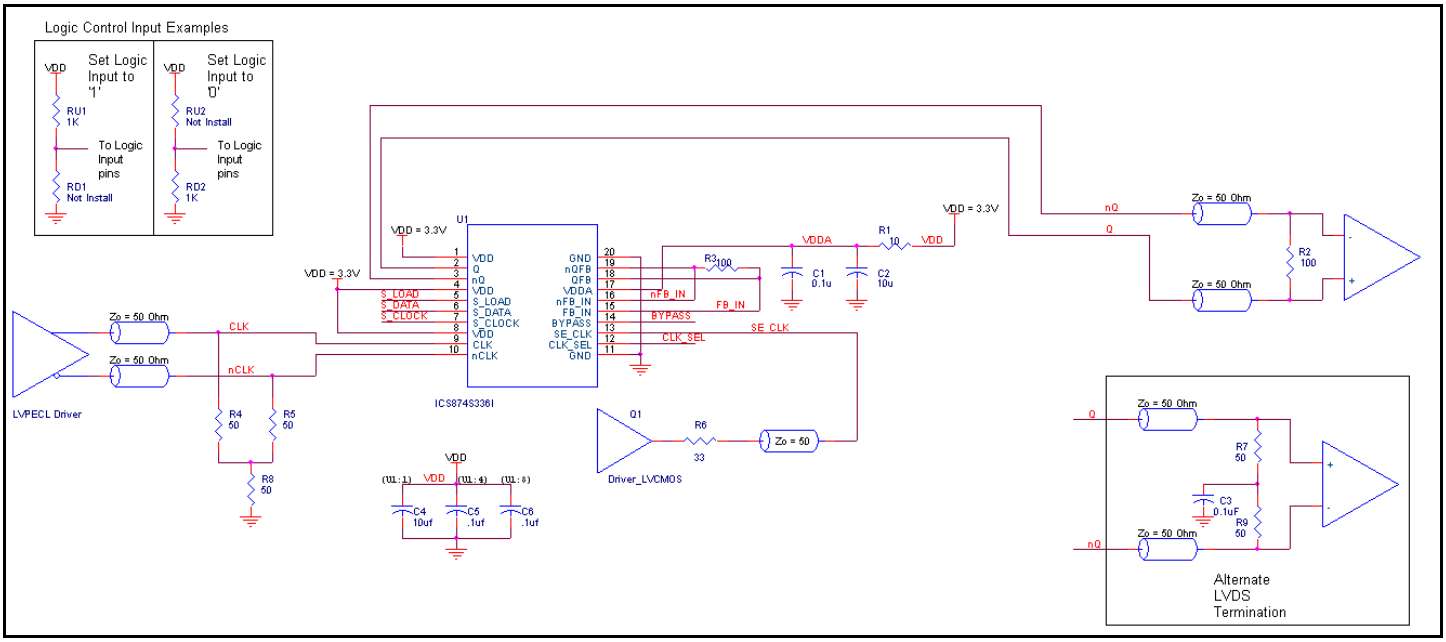



**Figure 6. ICS874S336 Schematic Example**

# Power Considerations

This section provides information on power dissipation and junction temperature for the ICS874S336.
Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the ICS74S336 is the sum of the core power plus the power dissipated in the load(s).
The following is the power dissipation for $V_{DD}$ = 3.3V + 5% = 3.465V, which gives worst case results.

- Power (core)$_{MAX}$ = $V_{DD\_MAX}$ * ($I_{DD\_MAX}$ + $I_{DDA\_MAX}$) = 3.465V * (115mA + 15mA) = **450.45mW**

### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS devices is 125°C.

The equation for Tj is as follows: Tj = $\theta_{JA}$ * Pd_total + $T_A$

Tj = Junction Temperature

$\theta_{JA}$ = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$ = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{JA}$ must be used. Assuming no air flow and a multi-layer board, the appropriate value is 87.2°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

70°C + 0.450W * 87.2°C/W = 109.2°C. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (single layer or multi-layer).

**Table 7. Thermal Resistance $\theta_{JA}$ for 20 Lead TSSOP, Forced Convection**

| $\theta_{JA}$ **by Velocity** | | | |
|---|---|---|---|
| Meters per Second | **0** | **1** | **2.5** |
| Multi-Layer PCB, JEDEC Standard Test Boards | 87.2°C/W | 82.9°C/W | 80.7°C/W |

# Reliability Information

**Table 8. $\theta_{JA}$ vs. Air Flow Table for a 20 Lead TSSOP**

| $\theta_{JA}$ by Velocity | | | |
|---|---|---|---|
| Meters per Second | **0** | **1** | **2.5** |
| Multi-Layer PCB, JEDEC Standard Test Boards | 87.2°C/W | 82.9°C/W | 80.7°C/W |

## Transistor Count

The transistor count for ICS874S336 is: 2434

# Package Outline and Package Dimension

**Package Outline - G Suffix for 20 Lead TSSOP**

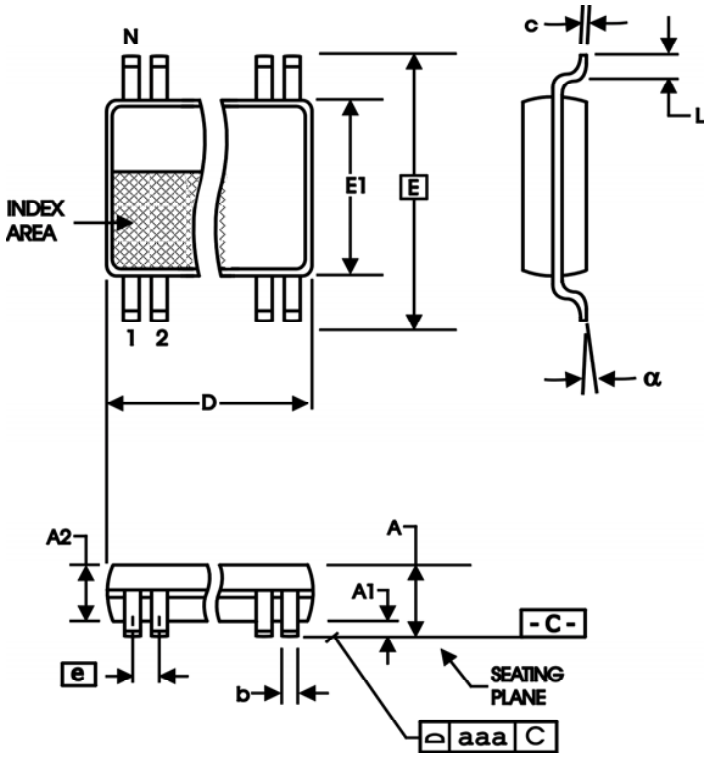



**Table 9. Package Dimensions**

| All Dimensions in Millimeters | | |
|---|---|---|
| **Symbol** | **Minimum** | **Maximum** |
| **N** | 20 | |
| **A** | | 1.20 |
| **A1** | 0.05 | 0.15 |
| **A2** | 0.80 | 1.05 |
| **b** | 0.19 | 0.30 |
| **c** | 0.09 | 0.20 |
| **D** | 6.40 | 6.60 |
| **E** | 6.40 Basic | |
| **E1** | 4.30 | 4.50 |
| **e** | 0.65 Basic | |
| **L** | 0.45 | 0.75 |
| $\alpha$ | 0° | 8° |
| **aaa** | | 0.10 |

Reference Document: JEDEC Publication 95, MO-153

# Ordering Information

**Table 9. Ordering Information**

| Part/Order Number | Marking | Package | Shipping Packaging | Temperature |
|---|---|---|---|---|
| 874S336AG | TBD | 20 Lead TSSOP | Tube | 0°C to 70°C |
| 874S336AGT | TBD | 20 Lead TSSOP | 2500 Tape & Reel | 0°C to 70°C |
| 874S336AGLF | ICS874S336AL | “Lead-Free” 20 Lead TSSOP | Tube | 0°C to 70°C |
| 874S336AGLFT | ICS874S336AL | “Lead-Free” 20 Lead TSSOP | 2500 Tape & Reel | 0°C to 70°C |

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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